

# 18V, 2A, Forced PWM Synchronous Step-Down Converter in SOT23-6

#### DESCRIPTION

ETA8121 is a wide input range, high-efficiency and high frequency DC-to-DC step-down switching regulator, capable of delivering up to 2A of output current. With a fixed switching frequency of 1MHz, this current mode PWM controlled converter allows the use of small external components, such as ceramic input and output caps, as well as small inductors. ETA8121 also employs a forced PWM control scheme, thereby ensuring the minimum ripple voltage. Its input can withstand a voltage surge as high as 24V, making it an ideal solution for industrial applications such as Surveillance Camera, LCD TV, Set Top Box, etc.

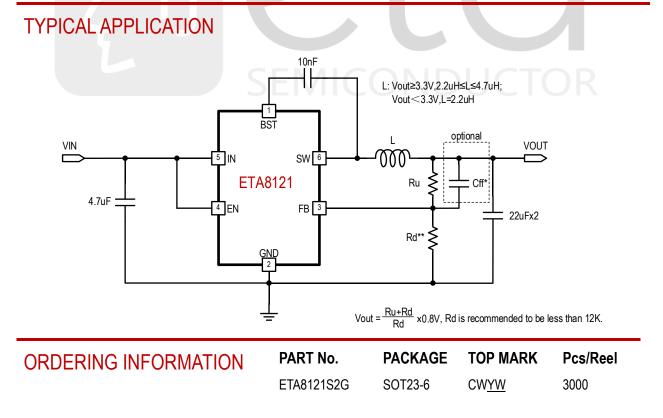
ETA8121 is available in SOT23-6 package.

#### FEATURES

- Wide Input Range: 4.2V-18V
- Forced PWM Control
- 1MHz Switching Frequency
- Capable of Delivering 2A Current
- No External Compensation Needed
- Current Mode Control
- Thermal Shutdown and UVLO
- Excellent Load and Line Transient Response
- Available in SOT23-6 Package
- RoHS Compliant
- Moisture Sensitivity Level 3

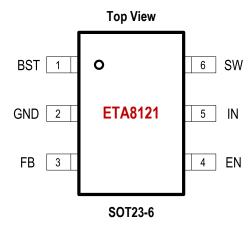
## **APPLICATIONS**

- Surveillance Camera
- Set Top Box
- LCD TV





### **PIN CONFIGURATION**



### ABSOLUTE MAXIMUM RATINGS

(Note: Exceeding these limits may damage the device. Exposure to absolute maximum rating conditions for long periods may affect device reliability.)

	-		
IN, EN Voltage		–(	).3V to 24V
SW Voltage0.3V (-5)	√ <10nS	) to 24V (	25V<10nS)
BST Voltage		–0.3\	/ to SW+6V
FB Voltage			-0.3V to 6V
Operating Temperature Ra	nge	40	°C to 85°C
Storage Temperature Rang	ge	–55°	C to 150°C
Thermal Resistance	$\theta_{JA}$	$\theta_{\text{JC}}$	
SOT23-6	180	90	°C/W
Lead Temperature (Solderi	ng 10se	c)	260°C

## ELECTRICAL CHARACTERISTICS

(V<sub>IN</sub> = 12V, V<sub>OUT</sub> = 3.3V, unless otherwise specified. Typical values are at TA =  $25^{\circ}$ C.)

PARAMETER	CONDITIONS	MIN	ТҮР	MAX	UNIT
Input Voltage Range		4.2		18	V
Input UVLO	Rising, Hysteresis=340mV		4.2		V
Input Supply Current	Vin=12V, lout=0, Vfb=0.85V, no switching		750		μA
Input Shutdown Current			7	14	μA
FB Voltage		0.784	0.8	0.816	V
FB Input Current			0	1	μA
Output Voltage Range	4.5V≤V <sub>IN</sub> ≤18V	0.8		12	V
Switching Frequency			1000		kHz
Maximum Duty Cycle			99		%
Short Circuit Hiccup Time	On Time		2		mS
Short Circuit Hiccup Time	Off Time		6		mS
FB Hiccup Threshold			0.2		V
High Side Switch On			160		mΩ
Low Side Switch On			95		mΩ
High Side Current Limit			3.5		А
SW Leakage Current	IN=SW=12V			40	μA
EN Rising Threshold		1.3	1.58	1.9	V
EN Falling Threshold		1.1	1.4	1.7	V
EN Input Current	V <sub>EN</sub> =2V		1		uA
Thermal Shutdown	Rising, Hysteresis =40°C		150		°C

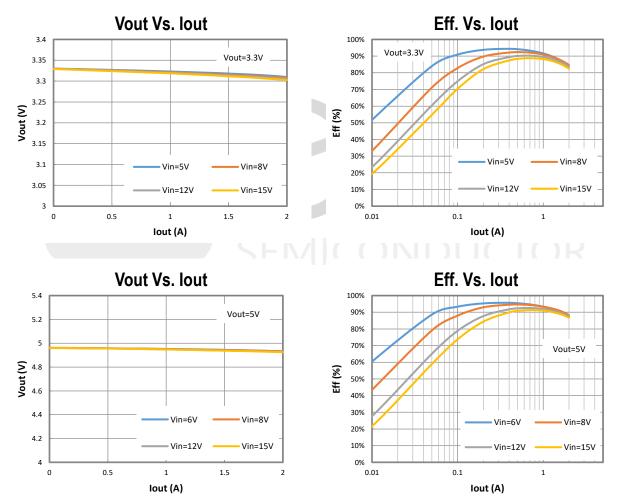


#### **PIN DESCRIPTION**

PIN #	NAME	DESCRIPTION
1	BST	Bootstrap pin. Connect a 10nF capacitor from this pin to SW
2	GND	Ground
3	FB	Feedback Input. Connect an external resistor divider from the output to FB and
		GND to set V <sub>OUT</sub>
4	EN	Enable pin for the IC. Drive this pin high to enable the part, low to disable.
5	VIN	Supply Voltage. Bypass with a 4.7µF ceramic capacitor to GND
6	SW	Inductor Connection. Connect two 22uF capacitors from this pin to GND

## TYPICAL CHARACTERISTICS

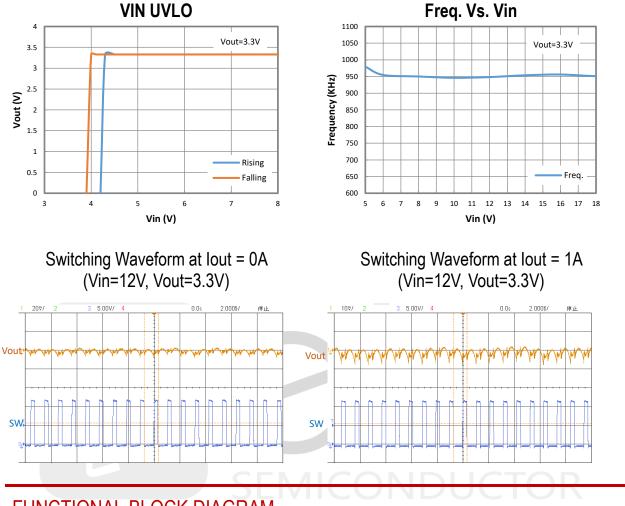
(Typical values are at  $T_A = 25^{\circ}C$  unless otherwise specified.)



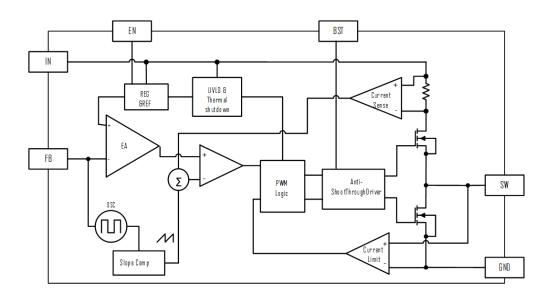


## TYPICAL CHARACTERISTICS

(Typical values are at T<sub>A</sub> = 25°C unless otherwise specified.)



## FUNCTIONAL BLOCK DIAGRAM





### FUNCTIONAL DESCRIPTION

The ETA8121 is a synchronous buck regulator ICs that integrates the PWM control, top and bottom switches on the same die to minimize the switching transition loss and conduction loss.

The ETA8121 is a wide input range, high efficiency, DC-to-DC step-down switching regulator, capable of delivering up to 2A of output current, integrated with a 160m $\Omega$  high side and 95m $\Omega$  low side MOSFET. It uses PWM current-mode control scheme. An error amplifier integrates error between the FB signal and the internal reference voltage. The output of the integrator is then compared to the sum of a current-sense signal and the slope compensation ramp. This operation generates a PWM signal that modulates the duty cycle of the power MOSFET to achieve regulation for output voltage.

#### Forced PWM Operation

A forced PWM DC-DC regulator always switches at a fixed frequency even when the output load is small. This is to ensure a minimum output voltage ripple over the full load range. It may induce a relatively large (about 10mA when Vin=12V) input current when load is zero.

#### Enable

EN is a digital control pin that turns the ETA8121 on and off. Drive EN High to turn on the regulator, drive it Low to turn it off. An internal  $1M\Omega$  resistor from EN pin to GND allows EN to float to shutdown the chip. Connecting the EN pin through a pull up resistor or shorted EN to IN will automatically turn on the chip whenever plug in IN.

#### Over Current Protection and Hiccup

ETA8121has a cycle-by-cycle over current limit for when the inductor current peak value is over the set current limit threshold. When the output voltage drop until FB falls below UV threshold (0.2V), the ETA8121 will enter hiccup mode. It will turn off the chip immediately for 6mS. After that, it will try to re-starts as normal for 2mS. After 2mS, if FB is still below UV threshold, then the chip enters hiccup mode again. If FB is higher than UV threshold, it will enter the normal mode.

#### Over-Temperature Protection

Thermal protection disables the output when the junction temperature rises to approximately 150°C, allowing the device to cool down. When the junction temperature cools to approximately 110°C, the output circuitry is again enabled. Depending on power dissipation, thermal resistance, and ambient temperature, the thermal protection circuit may cycle on and off. This cycling limits regulator dissipation, protecting the device from damage as a result of overheating.



### APPLICATION INFORMATION

#### External Output Voltage Setting

In external Output Voltage Setting Version selected, the ETA8121 regulator is programmed using an external resistor divider. The output voltage is calculated using below equation.

$$V_{OUT} = V_{REF} \times (1 + \frac{R_u}{R_d})$$

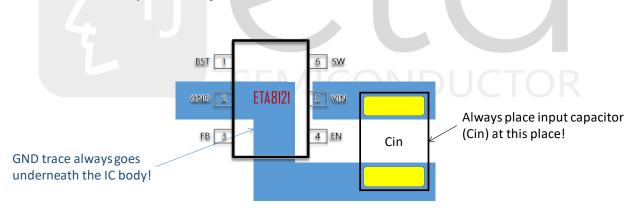
Where: VREF =0.8V typically (the internal reference voltage)

Resistors Rd has to be between 1kOhm to 12KOhm and thus Ru is calculated by following equation.

$$R_u = \left(\frac{V_{OUT}}{V_{REF}} - 1\right) \times R_d$$

### PCB LAYOUT GUIDE

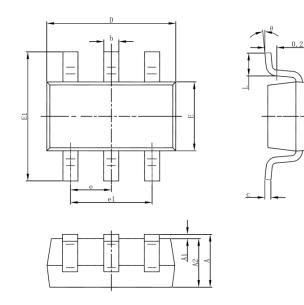
For any high voltage buck, it is always crucial to have input capacitor placed as close to the chip's IN and GND pin without any via, because the input capacitor is to keep the chip's real input voltage from dropping too much when large switching current is drawn from the input node. A simple illustration of how to place input capacitor and draw the trace to the chip's IN and GND pins is shown below, and it is highly recommended to strictly follow this guide.





# PACKAGE OUTLINE

### Package: SOT23-6



Cumb a I	Dimensions In Millimeters		Dimensions In Inches		
Symbol	Min	Max	Min	Max	
А	1.050	1.250	0.041	0.049	
A1	0.000	0.100	0.000	0.004	
A2	1.050	1.150	0.041	0.045	
b	0.300	0.500	0.012	0.020	
С	0.100	0.200	0.004	0.008	
D	2.820	3.020	0.111	0.119	
Е	1.500	1.700	0.059	0.067	
E1	2.650	2.950	0.104	0.116	
е	0.950	0.950(BSC)		(BSC)	
e1	1.800	2.000	0.071	0.079	
L	0.300	0.600	0.012	0.024	
θ	0°	8°	0°	8°	